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# FAN7387V

## Ballast Control IC for Compact Fluorescent Lamp

### Features

- Integrated Half-Bridge MOSFET
- Internal Clock Using RCT
- Enable External Sync Function Using RCT
- Dead-Time Control by using Resistor
- Shut Down (Disable Mode)
- Internal Shunt Regulator
- UVLO Function High and Low Side

### Applications

- Compact Fluorescent Lamp Ballast

### Description

The FAN7387V, developed using Fairchild's unique high-voltage process and system-in-package (SiP) concept, is a ballast-control integrated circuit (IC) for a compact fluorescent lamp (CFL). The FAN7387V has a simple oscillating circuit using an external resistor and capacitor so the frequency variation is stable across the temperature range. FAN7387V has a external pin for dead time control and shutdown. By using this resistor, a designer can choose the optimum dead time to reduce the power loss on internal switching devices (MOSFETs).

8-DIP



### Ordering Information

Part Number	Operating Temperature	Package	Packing Method
FAN7387VN	-40 to +125°C	8-Lead, Dual-In-line Package (DIP)	Tube



## Typical Applications Diagrams

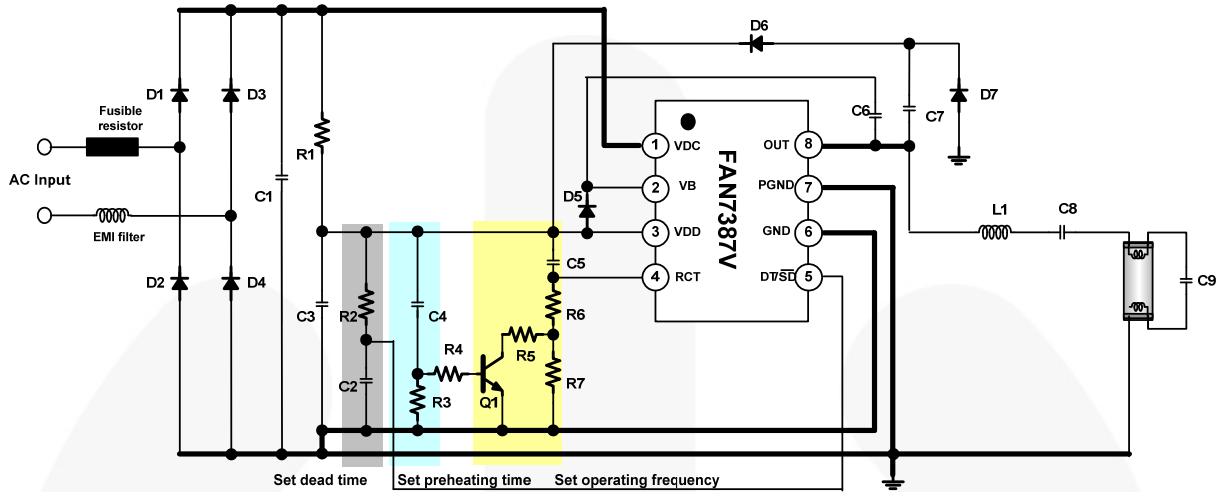


Figure 1. Typical Application Circuit for Fluorescent Lamp (Rapid Starting Method without PTC)

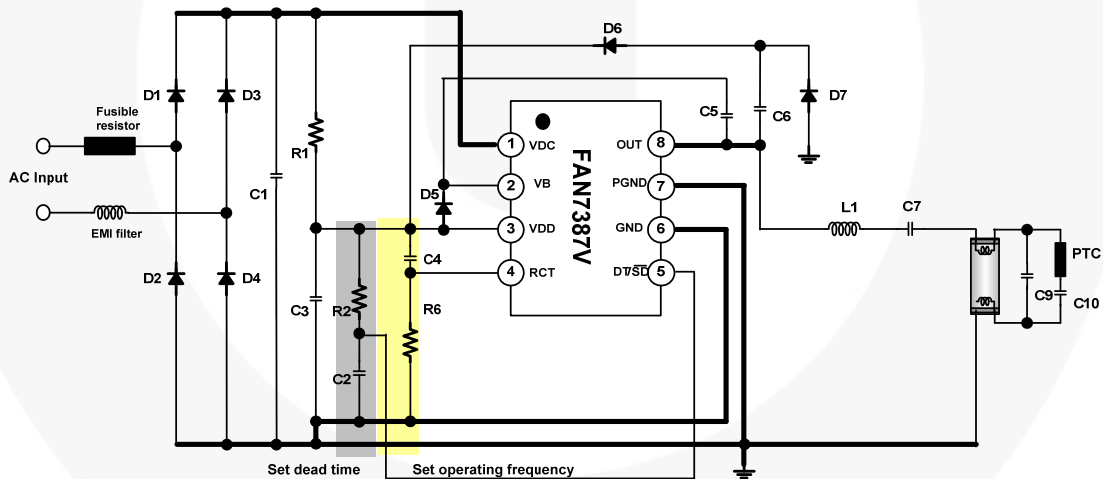


Figure 2. Typical Application Circuit for Fluorescent Lamp (Rapid Starting Method with PTC)

### Internal Block Diagram

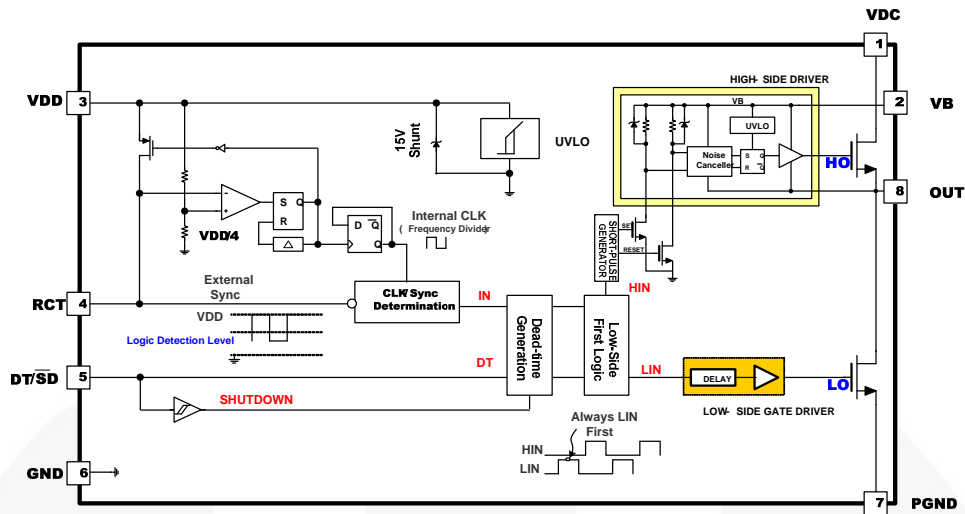


Figure 3. Functional Block Diagram

### Pin Configuration

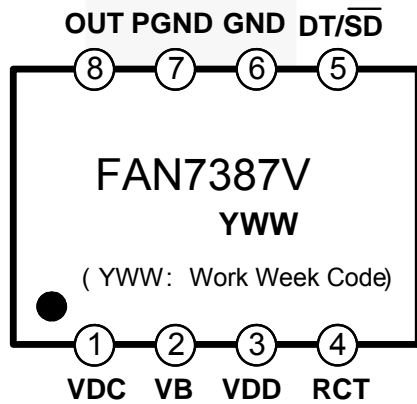


Figure 4. Pin Configurations (Top View)

### Pin Definitions

Pin #	Name	Description
1	VDC	High-voltage Supply
2	VB	High-Side Floating Supply
3	VDD	Supply Voltage
4	RCT	Oscillator Frequency Set Resistor and Capacitor
5	DT/SD	Dead Time Set Resistor
6	GND	Signal Ground
7	PGND	Power Ground
8	OUT	High-Side Floating Supply Return

## Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.  $T_A=25^{\circ}\text{C}$  unless otherwise specified.

Symbol	Parameter	Min.	Typ.	Max.	Unit
VB	High-Side Floating Supply	-0.3		465.0	V
V <sub>OUT</sub>	High-Side Floating Supply Return	-0.3		440.0	V
V <sub>RCT</sub>	RCT Pins Input Voltage		V <sub>DD</sub>		V
I <sub>CL</sub>	Clamping Current Level <sup>(1)</sup>			25	mA
dV <sub>OUT</sub> /dt	Allowable Offset Voltage Slew Rate		50		V/ns
T <sub>A</sub>	Operating Temperature Range	-40		+125	°C
T <sub>STG</sub>	Storage Temperature Range	-65		+150	°C
P <sub>D</sub>	Power Dissipation		2.1		W
Θ <sub>JA</sub>	Thermal Resistance (Junction-to-Air)		70		°C/W

### Note:

- Do not supply a low-impedance voltage source to the internal clamping Zener diode between the GND and the VDD pin of this device.

## Electrical Characteristics

$V_{BIAS}$  ( $V_{DD}$ ,  $V_B - V_{OUT}$ )=14.0V,  $T_A$ =25°C, unless otherwise specified.

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
<b>High Voltage Supply Section</b>						
$V_{DC}$	High Voltage Supply Voltage		440			V
<b>Low-Side Supply Characteristics (<math>V_{DD}</math>)</b>						
$V_{DDUV+}$	$V_{DD}$ UVLO Positive-Going Threshold	$V_{DD}$ Increasing	9	11	13	V
$V_{DDUV-}$	$V_{DD}$ UVLO Negative-Going Threshold	$V_{DD}$ Decreasing	7.8	8.8	9.8	
$V_{DDUHY}$	$V_{DD}$ -Side UVLO Hysteresis			2.2		
$V_{CL}$	Supply Camping Voltage	$I_{DD}=10mA$	14.4	15.4		
$I_{ST}$	Startup Supply Current	$V_{DD}=9V$		60	90	$\mu A$
$I_{QDD}$	Low-Side Quiescent Supply Current	$R_{DT}=100k\Omega$		230	380	
$I_{DD}$	Dynamic Operating Supply Current	20kHz, $C_L=1nF$		0.6		mA
<b>High-Side Supply Characteristics (<math>V_B - V_{OUT}</math>)</b>						
$V_{HSUV+}$	High-Side UVLO Positive-Going Threshold	$V_B - V_{OUT}$ Increasing	8	9	10	V
$V_{HSUV-}$	High-Side UVLO Negative-Going Threshold	$V_B - V_{OUT}$ Decreasing	7.5	8.5	9.5	
$V_{HSUHY}$	$V_{BS}$ Supply UVLO Hysteresis			0.5		
$I_{QHS}$	High-Side Quiescent Supply Current			50	90	$\mu A$
$I_{PBS}$	High-Side Dynamic Operating Supply Current	20kHz, $C_L=1nF$		130	180	
<b>Oscillator Characteristics</b>						
$f_{osc}$	Oscillation Frequency	$R_T=50k\Omega$ , $C_T=330pF$	18	20	22	kHz
D	Duty Cycle	Running Mode	47.5	49.0		%
$V_{RCT+}$	Upper Threshold Voltage of RCT	Running Mode		$V_{DD}$		V
$V_{RCT-}$	Lower Threshold Voltage of RCT	Running Mode		$V_{DD}/4$		
$V_{IH}$	Logic "1" Input Voltage of RCT	Running Mode		$3/4 V_{DD}$		
$V_{IL}$	Logic "0" Input Voltage of RCT	Running Mode			$3/5 V_{DD}$	
$t_D$	Dead Time	$R_{DT}=100k\Omega$	440	540	640	ns
$t_{DMIN}$	Minimum Dead Time	$V_{DT/\overline{SD}}=V_{DD}$	280	400	520	
<b>Protection Characteristics</b>						
$V_{SD+}$	Shutdown "1" Input Voltage	$V_{SD/\overline{DT}}=0$ After Run Mode	2.5			V
$V_{SD-}$	Shutdown "0" Input Voltage				1	
$I_{SD}$	Shutdown Current				350	$\mu A$
$t_{SD}$	Shutdown Propagation Delay			180		Ns
<b>Internal MOSFET Section</b>						
$I_{LK MOS}$	Internal MOSFET Leakage Current	$V_{DS}=400V$			50	$\mu A$
$R_{ON}$	Static Drain-Source On-Resistance	$V_{GS}=10V$ , $I_D=190mA$		4.6	6.0	$\Omega$
$I_S$	Maximum Continuous Drain-Source Diode Forward Current			0.38		A
$I_{SM}$	Maximum Pulsed Continuous Drain-Source Diode Forward Current			3.04		A
$V_{SD}$	Drain-Source Diode Forward Voltage	$V_{GS}=0V$ , $I_S=0.38A$			1.2	V

## Typical Performance Characteristics

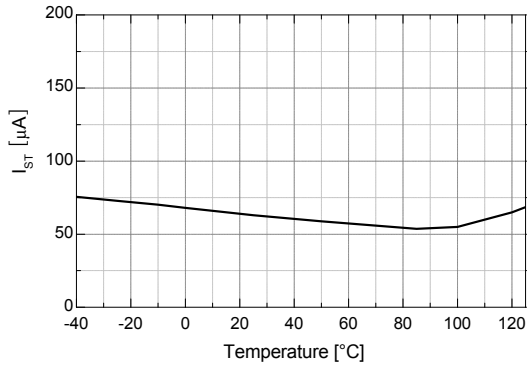


Figure 5. Startup Current vs. Temperature

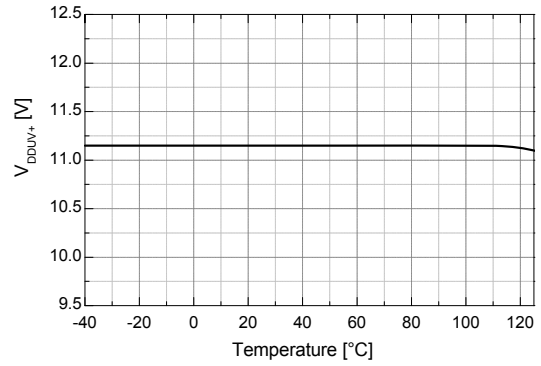


Figure 6.  $V_{DD}$  UVLO+ vs. Temperature

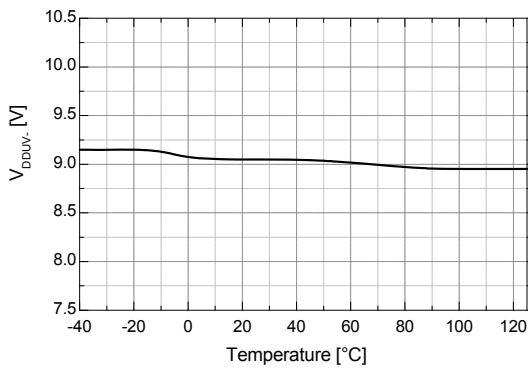


Figure 7.  $V_{DD}$  UVLO- vs. Temperature

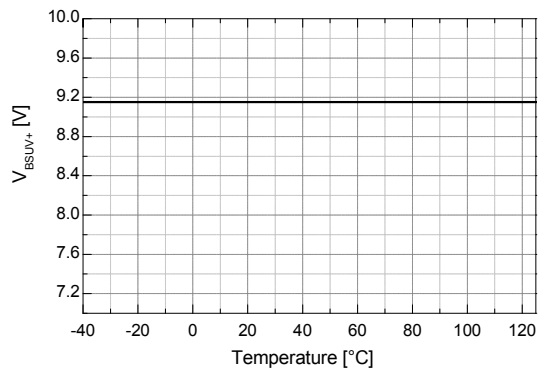


Figure 8.  $V_{B.V_{OUT}}$  UVLO+ vs. Temperature

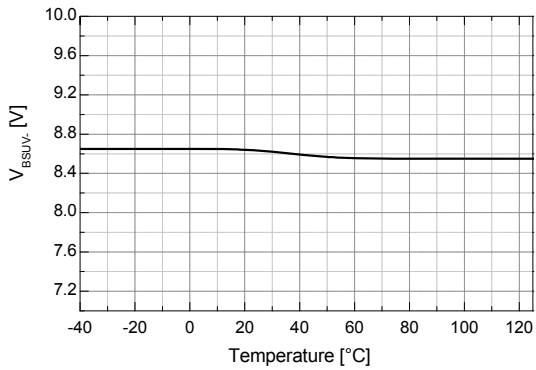


Figure 9.  $V_{B.V_{OUT}}$  UVLO- vs. Temperature

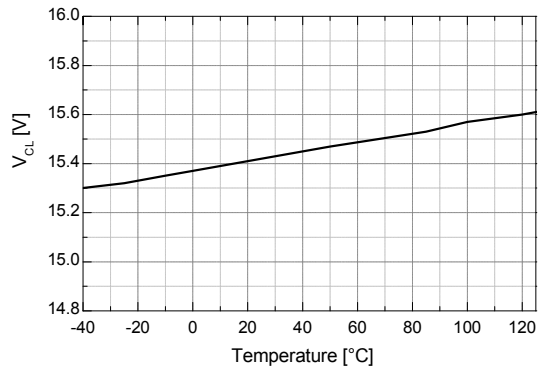
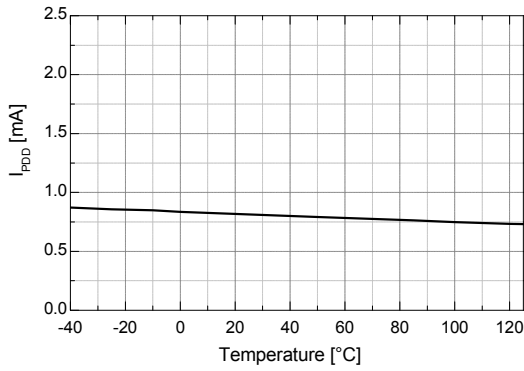
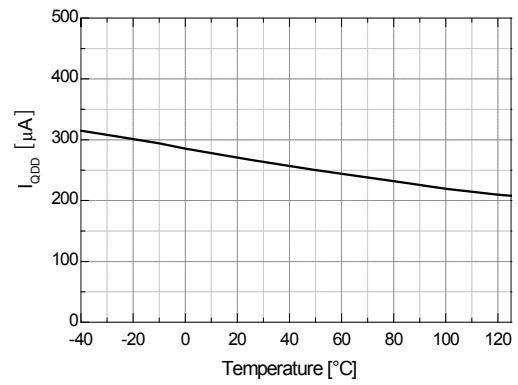


Figure 10.  $V_{CL}$  vs. Temperature

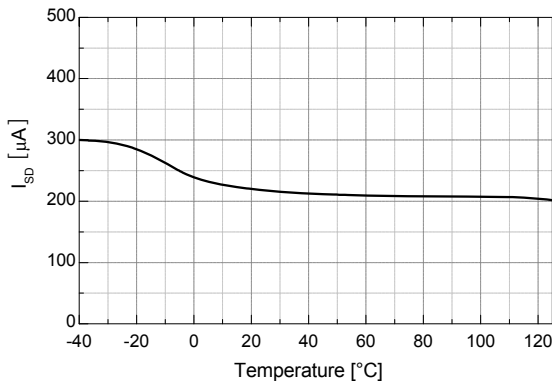
**Typical Performance Characteristics (Continued)**



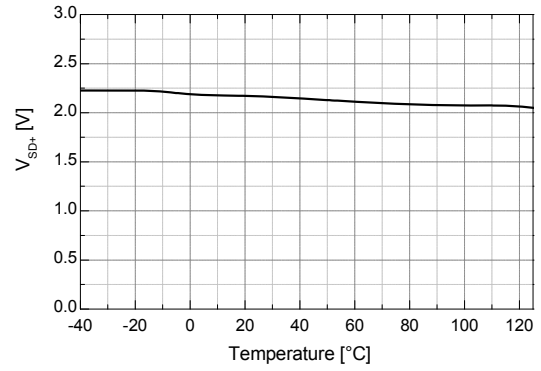
**Figure 11.  $I_{PDD}$  vs. Temperature**



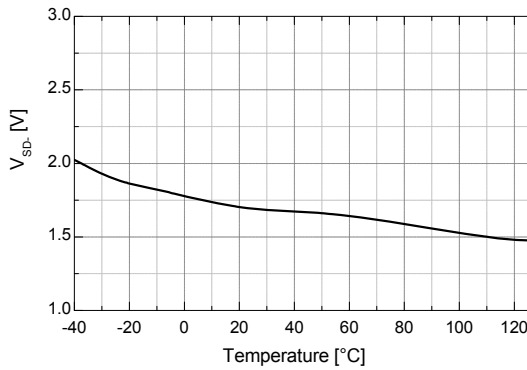
**Figure 12.  $I_{QDD}$  vs. Temperature**



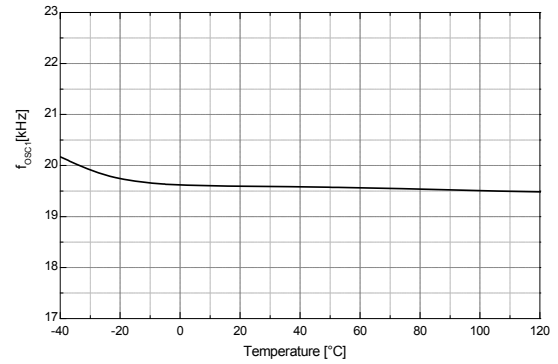
**Figure 13.  $I_{SD}$  vs. Temperature**



**Figure 14.  $V_{SD+}$  vs. Temperature**



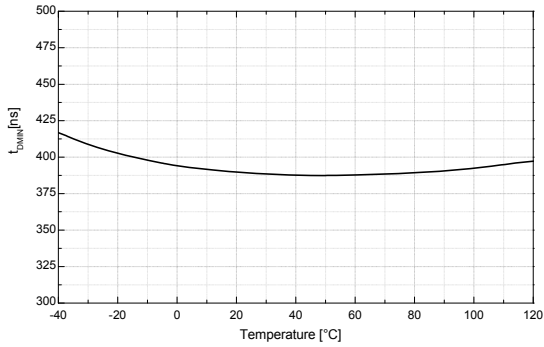
**Figure 15.  $V_{SD-}$  vs. Temperature**



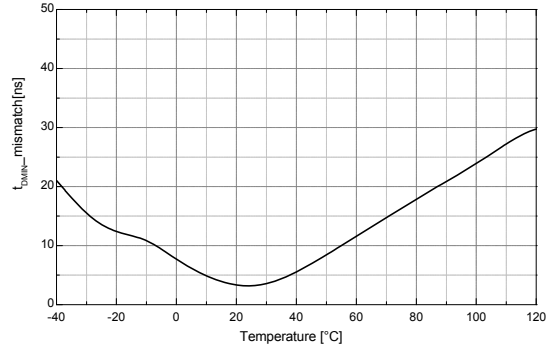
**Figure 16. Operating Frequency vs. Temperature**



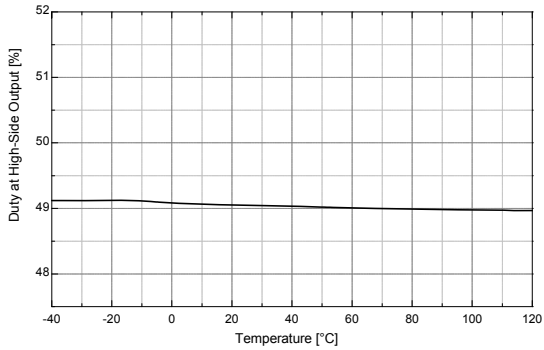
**Typical Performance Characteristics (Continued)**



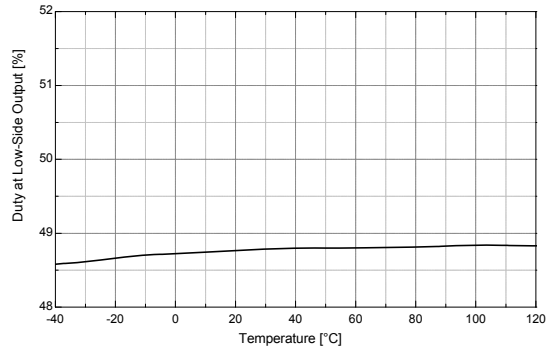
**Figure 17.  $t_{D\text{MIN}}$  vs. Temperature**



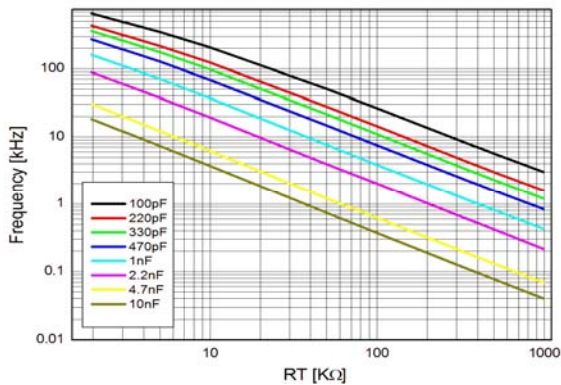
**Figure 18. Dead-Time Mismatch vs. Temperature**



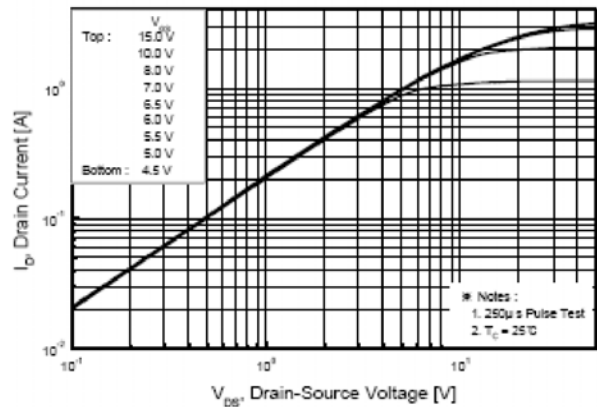
**Figure 19. High-Side Duty Ratio vs. Temperature**



**Figure 20. Low-Side Duty Ratio vs. Temperature**

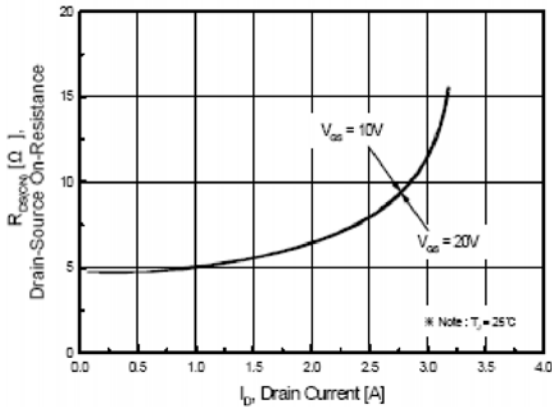


**Figure 21. Frequency vs.  $R_T$**

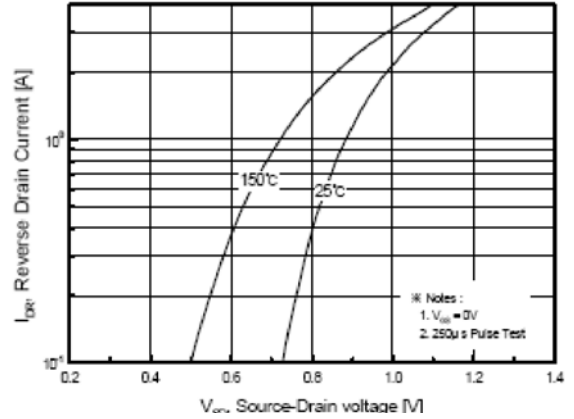


**Figure 22. On-Region Characteristics**

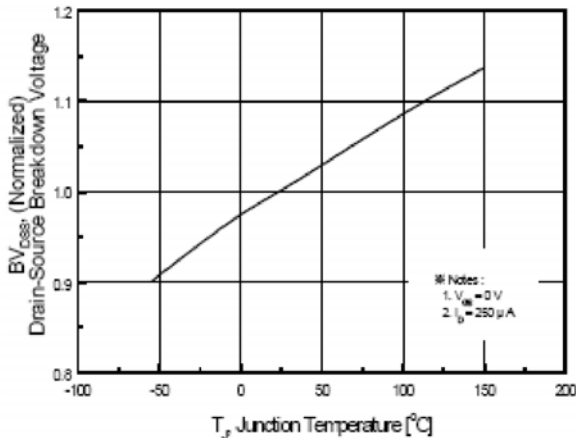
**Typical Performance Characteristics (Continued)**



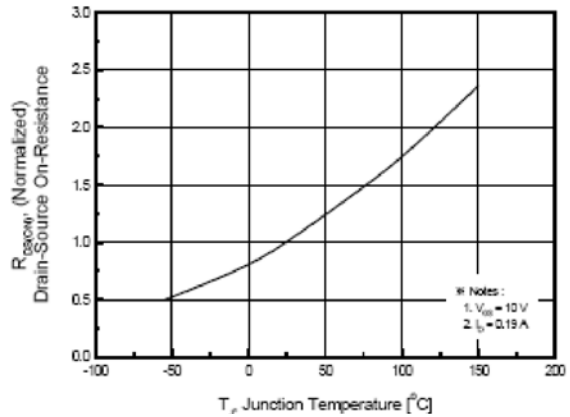
**Figure 23. On-Resistance Variation vs. Drain Current and Gate Voltage**



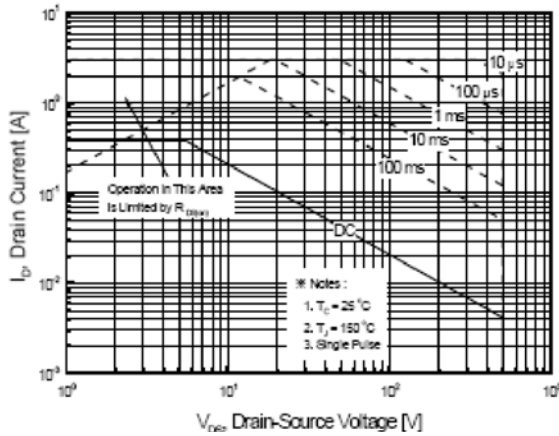
**Figure 24. Body Diode Forward Voltage Variation vs. Source Current and Temperature**



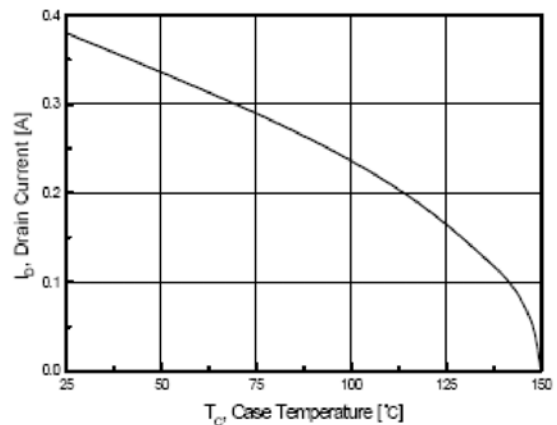
**Figure 25. Breakdown Voltage Variation vs. Temperature**



**Figure 26. On-Resistance Variation vs. Temperature**



**Figure 27. Maximum Safe Operating Area**



**Figure 28. Maximum Drain-Current vs. Case Temperature**

## Application Information

### 1. Under-Voltage Lockout (UVLO) Function

FAN7387V has a UVLO circuit for a low-side and high-side block. When  $V_{DD}$  reaches to the  $V_{DD_{UV+}}$ , the UVLO circuit is released and the FAN7387V operates normally. At UVLO condition, the FAN7387V has a low supply current of less than  $130\mu A$ . Once UVLO is released, FAN7387V operates normally until  $V_{DD}$  goes below  $V_{DD_{UV-}}$ , the UVLO hysteresis. FAN7387V also has a high-side gate driver. The supply for the high-side driver is applied between VB and  $V_{OUT}$ . To prevent malfunction at low supply voltage between VB and  $V_{OUT}$ , FAN7387V provides an additional UVLO circuit. If  $V_{B-V_{OUT}}$  is under  $V_{HSUV+}$ , the driver holds LOW state to turn off the high-side switch. Once the voltage of  $V_{B-V_{OUT}}$  is higher than  $V_{HSUV+}$ , after  $V_{B-V_{OUT}}$  exceeds  $V_{HSUV-}$ , the operation of driver resumes.

### 2. Oscillator

The running frequency is determined by an external timing resistor ( $R_T$ ) and timing capacitor ( $C_T$ ). The charge time of capacitor  $C_T$  from  $1/4 V_{DD}$  to  $V_{DD}$  determines the running frequency of gate driver output ( $V_{OUT}$ ).

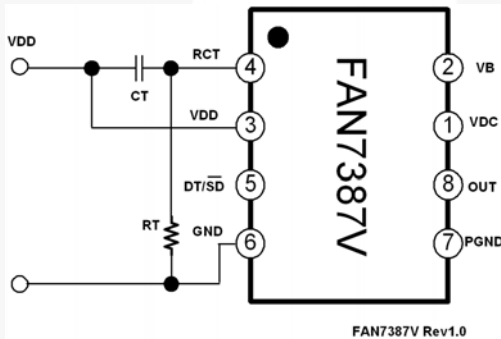


Figure 29. Typical Connection Method

Figure 30 shows the typical waveforms of RCT and internal signals (LO and HO) of IC. From the circuit analysis, the discharging time of RCT,  $t$ , is given by:

$$V_{RCT}(t) = V_{DD} \times \ln\left(\frac{-t}{R_T \cdot C_T}\right) \quad (1)$$

From Equation 1, it is possible to calculate the discharging time,  $t$ , from  $V_{DD}$  to one quarter ( $1/4$ ) of  $V_{DD}$  by substituting  $V_{RCT}(t)$  with  $1/4 V_{DD}$ .

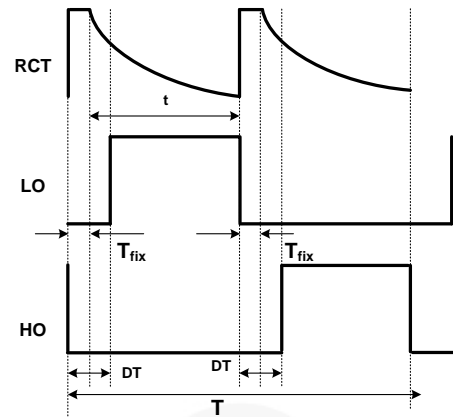


Figure 30. Typical Waveforms of RCT and Internal Signal (LO, HO) of IC

$$t = 1.38 \cdot R_T \cdot C_T \quad (2)$$

The running frequency of IC is determined by  $1/t$  and is approximately given as:

$$f_{\text{running}} = \frac{1}{t} = \frac{1}{2(t + t_{\text{fix}})} \quad (3)$$

where  $t$  is the discharging time of the RCT voltage and  $t_{\text{fix}}$  is constant value about 450ns of IC.

### 3. Programming Dead Time Control / Shutdown

A multi-function pin controls dead time using an external resistor ( $R_{DT}$ ) and protects abnormal condition using an external switch. This pin should be connected to an external capacitor to maintain stable operation.

If the voltage of DT/SD is decreased to under 1V by an external switch, such as the TR or MOSFET, the FAN7387V enters shutdown mode. In this mode, the FAN7387V doesn't have any output signal.

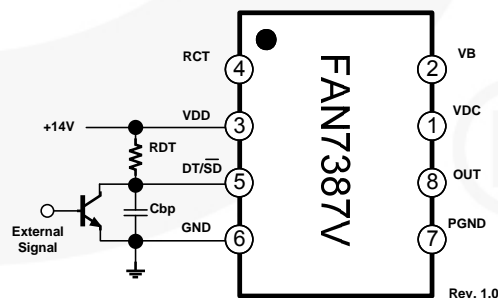
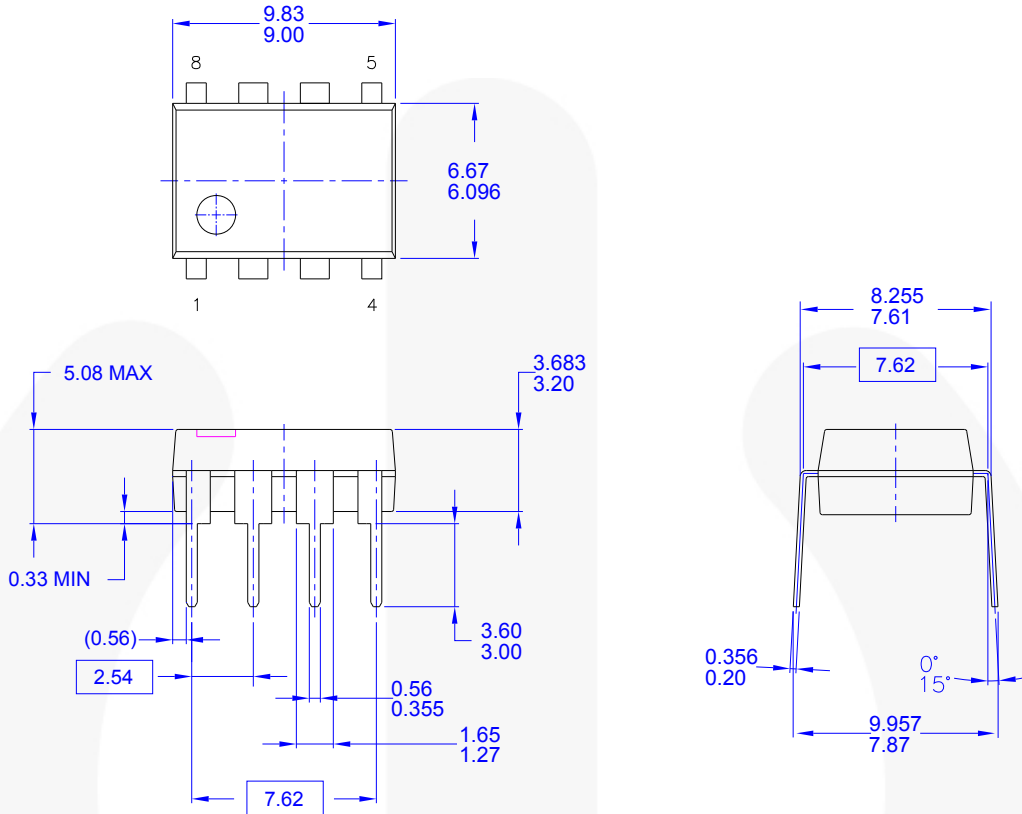


Figure 31. External Shutdown Circuit

## Physical Dimensions



- NOTES: UNLESS OTHERWISE SPECIFIED
- THIS PACKAGE CONFORMS TO JEDEC MS-001 VARIATION BA
  - ALL DIMENSIONS ARE IN MILLIMETERS.
  - DIMENSIONS ARE EXCLUSIVE OF BURRS, MOLD FLASH, AND TIE BAR EXTRUSIONS.
  - DIMENSIONS AND TOLERANCES PER ASME Y14.5M-1994
  - DRAWING FILENAME AND REVISION: MKT-N08FRE2.

**Figure 32. 8-Lead Dual Inline Package (DIP)**



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| BitSiC™  | GreenBridge™                                   | QFET®  | TinyBuck™   |
| Build it Now™  | Green FPS™                                     | QS™  | TinyCalc™   |
| CorePLUS™  | Green FPS™ e-Series™                           | Quiet Series™  | TinyLogic®  |
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| CROSSVOLT™   | GTO™   |  ™                | TinyPower™  |
| CTL™   | IntelliMAX™                                    | Saving our world, 1mW/W/kW at a time™  | TinyPWM™  |
| Current Transfer Logic™  | ISOPLANAR™                                     | SignalWise™  | TinyWire™   |
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| FAST®  | mWSaver™                                       | SupreMOS®  | VisualMax™  |
| FastvCore™   | OptoHiT™                                       | SyncFET™   | VoltagePlus™  |
| FETBench™  | OPTOLOGIC®                                     | Sync-Lock™   | XS™   |
| FlashWriter®*  | OPTOPLANAR®                                    |  SYSTEM GENERAL®* |   |
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**Definition of Terms**

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Rev. I61

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